10MHz, 6V/µs, Dual/Quad Rail-to-Rail Input and Output Precision C-Load Op Amps

FEATURES

- **Rail-to-Rail Input and Output**
- **475µV Max V_{OS} from V^+ to V^-**
- **Gain-Bandwidth Product: 10MHz**
- **Slew Rate: 6V/µs**
- **Low Supply Current per Amplifier: 1.7mA**
- Input Offset Current: 65nA Max
- Input Bias Current: 650nA Max
- Open-Loop Gain: 1000V/mV Min
- Low Input Noise Voltage: 12nV/√Hz Typ
- Wide Supply Range: 2.2V to ±15V
- Large Output Drive Current: 30mA
- Stable for Capacitive Loads Up to 10,000pF
- Dual in 8-Pin PDIP and SO Package
- Quad in Narrow 14-Pin SO

APPLICATIONS

- Driving A-to-D Converters
- Active Filters
- Rail-to-Rail Buffer Amplifiers
- Low Voltage Signal Processing
- Battery-Powered Systems

DESCRIPTION

The LT®1498/LT1499 are dual/quad, rail-to-rail input and output precision C-Load™ op amps with a 10MHz gain-bandwidth product and a 6V/µs slew rate.

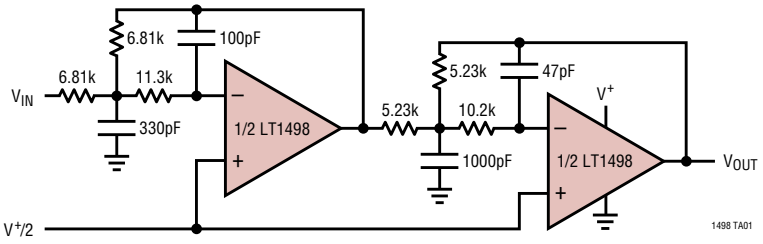
The LT1498/LT1499 are designed to maximize input dynamic range by delivering precision performance over the full supply voltage. Using a patented technique, both input stages of the LT1498/LT1499 are trimmed, one at the negative supply and the other at the positive supply. The resulting guaranteed common mode rejection is much better than other rail-to-rail input op amps. When used as a unity-gain buffer in front of single supply 12-bit A-to-D converters, the LT1498/LT1499 are guaranteed to add less than 1LSB of error even in single 3V supply systems.

With 110dB of supply rejection, the LT1498/LT1499 maintain their performance over a supply range of 2.2V to 36V and are specified for 3V, 5V and ±15V supplies. The inputs can be driven beyond the supplies without damage or phase reversal of the output. These op amps remain stable while driving capacitive loads up to 10,000pF.

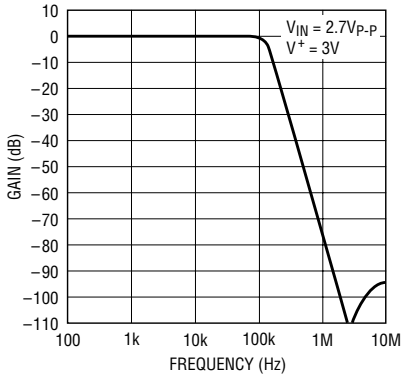
The LT1498 is available with the standard dual op amp configuration in 8-pin PDIP and SO packaging. The LT1499 features the standard quad op amp configuration and is available in a 14-pin plastic SO package. These devices can be used as plug-in replacements for many standard op amps to improve input/output range and precision.

TYPICAL APPLICATION

Single Supply 100kHz 4th Order Butterworth Filter



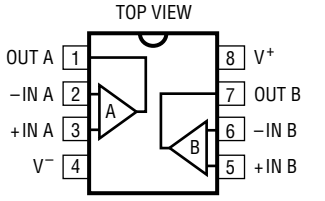
Frequency Response



ABSOLUTE MAXIMUM RATINGS (Note 1)

Total Supply Voltage (V^+ to V^-) 36V
Input Current ..... ±10mA
Output Short-Circuit Duration (Note 2) ..... Continuous
Operating Temperature Range ..... –40°C to 85°C
Specified Temperature Range (Note 4) .... –40°C to 85°C
Junction Temperature ..... 150°C
Storage Temperature Range ..... –65°C to 150°C
Lead Temperature (Soldering, 10 sec) ..... 300°C

PACKAGE/ORDER INFORMATION



N8 PACKAGE
8-LEAD PDIP

S8 PACKAGE
8-LEAD PLASTIC SO

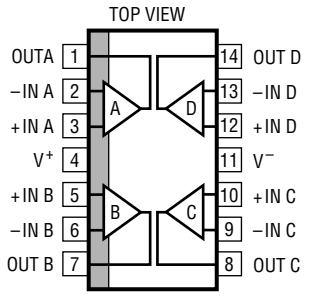
T_{JMAX} = 150°C, θ_{JA} = 130°C/W (N8)
T_{JMAX} = 150°C, θ_{JA} = 190°C/W (S8)

ORDER PART NUMBER

LT1498CN8
LT1498CS8
LT1498IN8
LT1498IS8

S8 PART MARKING

1498
1498I



S PACKAGE
14-LEAD PLASTIC SO

T_{JMAX} = 150°C, θ_{JA} = 150°C/W

ORDER PART NUMBER

LT1499CS
LT1499IS

Consult LTC Marketing for parts specified with wider operating temperature ranges.

ELECTRICAL CHARACTERISTICS

T_A = 25°C, V_S = 5V,0V; V_S = 3V,0V; V_{CM} = V_{OUT} = half supply, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	$V_{CM} = V^+$ $V_{CM} = V^-$		150 150	475 475	µV µV
ΔV_{OS}	Input Offset Voltage Shift	$V_{CM} = V^-$ to V^+		150	425	µV
	Input Offset Voltage Match (Channel-to-Channel)	$V_{CM} = V^+$, V^- (Note 5)		200	750	µV
I_B	Input Bias Current	$V_{CM} = V^+$ $V_{CM} = V^-$	0 –650	250 –250	650 0	nA nA
ΔI_B	Input Bias Current Shift	$V_{CM} = V^-$ to V^+		500	1300	nA
	Input Bias Current Match (Channel-to-Channel)	$V_{CM} = V^+$ (Note 5) $V_{CM} = V^-$ (Note 5)	0 –100	10 –10	100 0	nA nA
I_{OS}	Input Offset Current	$V_{CM} = V^+$ $V_{CM} = V^-$		5 5	65 65	nA nA
ΔI_{OS}	Input Offset Current Shift	$V_{CM} = V^-$ to V^+		10	130	nA
	Input Noise Voltage	0.1Hz to 10Hz		400		nV_{P-P}
e_n	Input Noise Voltage Density	f = 1kHz		12		$nV/\sqrt{Hz}$
i_n	Input Noise Current Density	f = 1kHz		0.3		$pA/\sqrt{Hz}$
C_{IN}	Input Capacitance			5		pF
A_{VOL}	Large-Signal Voltage Gain	V_S = 5V, V_O = 75mV to 4.8V, R_L = 10k V_S = 3V, V_O = 75mV to 2.8V, R_L = 10k	600 500	3800 2000		V/mV V/mV

ELECTRICAL CHARACTERISTICS

V_S = 5V,0V; V_S = 3V,0V; V_{CM} = V_{OUT} = half supply, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
CMRR	Common Mode Rejection Ratio	V_S = 5V, V_{CM} = V^- to V^+ V_S = 3V, V_{CM} = V^- to V^+	81 76	90 86		dB dB
	CMRR Match (Channel-to-Channel) (Note 5)	V_S = 5V, V_{CM} = V^- to V^+ V_S = 3V, V_{CM} = V^- to V^+	75 70	91 86		dB dB
PSRR	Power Supply Rejection Ratio	V_S = 2.2V to 12V, V_{CM} = V_O = 0.5V	88	105		dB
	PSRR Match (Channel-to-Channel) (Note 5)	V_S = 2.2V to 12V, V_{CM} = V_O = 0.5V	82	103		dB
V_{OL}	Output Voltage Swing (Low) (Note 6)	No Load I_{SINK} = 0.5mA I_{SINK} = 2.5mA		14 35 90	30 70 200	mV mV mV
V_{OH}	Output Voltage Swing (High) (Note 6)	No Load I_{SOURCE} = 0.5mA I_{SOURCE} = 2.5mA		2.5 50 140	10 100 250	mV mV mV
I_{SC}	Short-Circuit Current	V_S = 5V V_S = 3V	±12.5 ±12.0	±24 ±19		mA mA
I_S	Supply Current per Amplifier			1.7	2.2	mA
GBW	Gain-Bandwidth Product (Note 7)		6.8	10.5		MHz
SR	Slew Rate (Note 8)	V_S = 5V, A_V = –1, R_L = Open, V_O = 4V V_S = 3V, A_V = –1, R_L = Open	2.6 2.3	4.5 4.0		V/µs V/µs

The ● denotes the specifications which apply over the temperature range 0°C < T_A < 70°C. V_S = 5V, 0V; V_S = 3V, 0V; V_{CM} = V_{OUT} = half supply, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	V_{CM} = V^+ V_{CM} = V^- + 0.1V	● ●		175 175	650 650	µV µV
V_{OS} TC	Input Offset Voltage Drift (Note 3)	 V_{CM} = V^+	● ●		0.5 1.5	2.5 4.0	µV/°C µV/°C
ΔV_{OS}	Input Offset Voltage Shift	V_{CM} = V^- + 0.1V to V^+	●		170	600	µV
	Input Offset Voltage Match (Channel-to-Channel)	V_{CM} = V^- + 0.1V, V^+ (Note 5)	●		200	900	µV
I_B	Input Bias Current	V_{CM} = V^+ V_{CM} = V^- + 0.1V	● ●	0 –780	275 –275	780 0	nA nA
ΔI_B	Input Bias Current Shift	V_{CM} = V^- + 0.1V to V^+	●		550	1560	nA
	Input Bias Current Match (Channel-to-Channel)	V_{CM} = V^+ (Note 5) V_{CM} = V^- + 0.1V (Note 5)	● ●	0 –170	15 –15	170 0	nA nA
I_{OS}	Input Offset Current	V_{CM} = V^+ V_{CM} = V^- + 0.1V	● ●		10 10	85 85	nA nA
ΔI_{OS}	Input Offset Current Shift	V_{CM} = V^- + 0.1V to V^+	●		20	170	nA
A_{VOL}	Large-Signal Voltage Gain	V_S = 5V, V_O = 75mV to 4.8V, R_L = 10k V_S = 3V, V_O = 75mV to 2.8V, R_L = 10k	● ●	500 400	2500 2000		V/mV V/mV
CMRR	Common Mode Rejection Ratio	V_S = 5V, V_{CM} = V^- + 0.1V to V^+ V_S = 3V, V_{CM} = V^- + 0.1V to V^+	● ●	78 73	89 85		dB dB
	CMRR Match (Channel-to-Channel) (Note 5)	V_S = 5V, V_{CM} = V^- + 0.1V to V^+ V_S = 3V, V_{CM} = V^- + 0.1V to V^+	● ●	74 69	90 86		dB dB
PSRR	Power Supply Rejection Ratio	V_S = 2.3V to 12V, V_{CM} = V_O = 0.5V	●	86	102		dB
	PSRR Match (Channel-to-Channel) (Note 5)	V_S = 2.3V to 12V, V_{CM} = V_O = 0.5V	●	80	102		dB

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the temperature range $0°C < T_A < 70°C$. V_S = 5V, 0V; V_S = 3V, 0V; $V_{CM} = V_{OUT}$ = half supply, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
V_{OL}	Output Voltage Swing (Low) (Note 6)	No Load	●		17	35	mV
		I_{SINK} = 0.5mA	●		40	80	mV
		I_{SINK} = 2.5mA	●		110	220	mV
V_{OH}	Output Voltage Swing (High) (Note 6)	No Load	●		3.5	15	mV
		I_{SOURCE} = 0.5mA	●		55	120	mV
		I_{SOURCE} = 2.5mA	●		160	300	mV
I_{SC}	Short-Circuit Current	V_S = 5V	●	±12	±23		mA
		V_S = 3V	●	±10	±20		mA
I_S	Supply Current per Amplifier		●		1.9	2.6	mA
GBW	Gain-Bandwidth Product (Note 7)		●	6.1	9		MHz
SR	Slew Rate (Note 8)	V_S = 5V, A_V = –1, R_L = Open, V_O = 4V	●	2.5	4.0		V/µs
		V_S = 3V, A_V = –1, R_L = Open	●	2.2	3.5		V/µs

The ● denotes the specifications which apply over the temperature range $-40°C < T_A < 85°C$. V_S = 5V, 0V; V_S = 3V, 0V; $V_{CM} = V_{OUT}$ = half supply, unless otherwise noted. (Note 4)

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	$V_{CM} = V^+$	●		250	750	µV
		$V_{CM} = V^- + 0.1V$	●		250	750	µV
V_{OS} TC	Input Offset Voltage Drift (Note 3)		●		0.5	2.5	µV/°C
		$V_{CM} = V^+$	●		1.5	4.0	µV/°C
ΔV_{OS}	Input Offset Voltage Shift	$V_{CM} = V^- + 0.1V$ to V^+	●		250	650	µV
	Input Offset Voltage Match (Channel-to-Channel)	$V_{CM} = V^- + 0.1V$, V^+ (Note 5)	●		300	1500	µV
I_B	Input Bias Current	$V_{CM} = V^+$	●	0	350	975	nA
		$V_{CM} = V^- + 0.1V$	●	–975	–350	0	nA
ΔI_B	Input Bias Current Shift	$V_{CM} = V^- + 0.1V$ to V^+	●		700	1950	nA
	Input Bias Current Match (Channel-to-Channel)	$V_{CM} = V^+$ (Note 5)	●	0	30	180	nA
		$V_{CM} = V^- + 0.1V$ (Note 5)	●	–180	–30	0	nA
I_{OS}	Input Offset Current	$V_{CM} = V^+$	●		15	110	nA
		$V_{CM} = V^- + 0.1V$	●		15	110	nA
ΔI_{OS}	Input Offset Current Shift	$V_{CM} = V^- + 0.1V$ to V^+	●		30	220	nA
A_{VOL}	Large-Signal Voltage Gain	V_S = 5V, V_O = 75mV to 4.8V, R_L = 10k	●	400	2500		V/mV
		V_S = 3V, V_O = 75mV to 2.8V, R_L = 10k	●	300	2000		V/mV
CMRR	Common Mode Rejection Ratio	V_S = 5V, $V_{CM} = V^- + 0.1V$ to V^+	●	77	86		dB
		V_S = 3V, $V_{CM} = V^- + 0.1V$ to V^+	●	73	81		dB
	CMRR Match (Channel-to-Channel) (Note 5)	V_S = 5V, $V_{CM} = V^- + 0.1V$ to V^+	●	72	86		dB
		V_S = 3V, $V_{CM} = V^- + 0.1V$ to V^+	●	69	83		dB
PSRR	Power Supply Rejection Ratio	V_S = 2.5V to 12V, $V_{CM} = V_O$ = 0.5V	●	86	100		dB
	PSRR Match (Channel-to-Channel) (Note 5)	V_S = 2.5V to 12V, $V_{CM} = V_O$ = 0.5V	●	80	100		dB
V_{OL}	Output Voltage Swing (Low) (Note 6)	No Load	●		18	40	mV
		I_{SINK} = 0.5mA	●		45	80	mV
		I_{SINK} = 2.5mA	●		110	220	mV
V_{OH}	Output Voltage Swing (High) (Note 6)	No Load	●		3.5	15	mV
		I_{SOURCE} = 0.5mA	●		60	120	mV
		I_{SOURCE} = 2.5mA	●		170	300	mV

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the temperature range $-40°C < T_A < 85°C$. $V_S = 5V, 0V$; $V_S = 3V, 0V$; $V_{CM} = V_{OUT}$ = half supply, unless otherwise noted. (Note 4)

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
I_{SC}	Short-Circuit Current	$V_S = 5V$ $V_S = 3V$	● ●	±7.5 ±7.5	±15 ±15		mA mA
I_S	Supply Current per Amplifier		●		2.0	2.7	mA
GBW	Gain-Bandwidth Product (Note 7)		●	5.8	8.5		MHz
SR	Slew Rate (Note 8)	$V_S = 5V$, $A_V = -1$, R_L = Open, $V_O = 4V$ $V_S = 3V$, $A_V = -1$, R_L = Open	● ●	2.2 1.9	3.6 3.2		V/µs V/µs

$T_A = 25°C$, $V_S = \pm 15V$, $V_{CM} = 0V$, $V_{OUT} = 0V$, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	$V_{CM} = V^+$ $V_{CM} = V^-$		200 200	800 800	µV µV
ΔV_{OS}	Input Offset Voltage Shift	$V_{CM} = V^-$ to V^+		150	650	µV
	Input Offset Voltage Match (Channel-to-Channel)	$V_{CM} = V^+, V^-$ (Note 5)		250	1400	µV
I_B	Input Bias Current	$V_{CM} = V^+$ $V_{CM} = V^-$	0 –715	250 –250	715 0	nA nA
ΔI_B	Input Bias Current Shift	$V_{CM} = V^-$ to V^+		500	1430	nA
	Input Bias Current Match (Channel-to-Channel)	$V_{CM} = V^+$ (Note 5) $V_{CM} = V^-$ (Note 5)	0 –120	12 –12	120 0	nA nA
I_{OS}	Input Offset Current	$V_{CM} = V^+$ $V_{CM} = V^-$		6 6	70 70	nA nA
ΔI_{OS}	Input Offset Current Shift	$V_{CM} = V^-$ to V^+		12	140	nA
	Input Noise Voltage	0.1Hz to 10Hz		400		nV_{P-P}
e_n	Input Noise Voltage Density	f = 1kHz		12		$nV/\sqrt{Hz}$
i_n	Input Noise Current Density	f = 1kHz		0.3		$pA/\sqrt{Hz}$
A_{VOL}	Large-Signal Voltage Gain	$V_O = -14.5V$ to 14.5V, $R_L = 10k$ $V_O = -10V$ to 10V, $R_L = 2k$	1000 500	5200 2300		V/mV V/mV
	Channel Separation	$V_O = -10V$ to 10V, $R_L = 2k$	116	130		dB
CMRR	Common-Mode Rejection Ratio	$V_{CM} = V^-$ to V^+	93	106		dB
	CMRR Match (Channel-to-Channel) (Note 5)	$V_{CM} = V^-$ to V^+	87	103		dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 5V$ to $\pm 15V$	89	110		dB
	PSRR Match (Channel-to-Channel) (Note 5)	$V_S = \pm 5V$ to $\pm 15V$	83	105		dB
V_{OL}	Output Voltage Swing (Low) (Note 6)	No Load I_{SINK} = 0.5mA I_{SINK} = 10mA		18 40 230	30 80 500	mV mV mV
V_{OH}	Output Voltage Swing (High) (Note 6)	No Load I_{SINK} = 0.5mA I_{SINK} = 10mA		2.5 55 420	10 120 800	mV mV mV
I_{SC}	Short-Circuit Current		±15	±30		mA
I_S	Supply Current per Amplifier			1.8	2.5	mA
GBW	Gain-Bandwidth Product (Note 7)		6.8	10.5		MHz
SR	Slew Rate	$A_V = -1$, R_L = Open, $V_O = \pm 10V$ Measure at $V_O = \pm 5V$	3.5	6		V/µs

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the temperature range $0°C < T_A < 70°C$. $V_S = \pm 15V$, $V_{CM} = 0V$, $V_{OUT} = 0V$, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	$V_{CM} = V^+$	●		200	900	µV
		$V_{CM} = V^- + 0.1V$	●		200	900	µV
V_{OS} TC	Input Offset Voltage Drift (Note 3)		●		1.0	3.5	µV/°C
		$V_{CM} = V^+$	●		2.0	5.0	µV/°C
ΔV_{OS}	Input Offset Voltage Shift	$V_{CM} = V^- + 0.1V$ to V^+	●		200	750	µV
	Input Offset Voltage Match (Channel-to-Channel)	$V_{CM} = V^- + 0.1V$, V^+ (Note 5)	●		350	1500	µV
I_B	Input Bias Current	$V_{CM} = V^+$	●	0	300	875	nA
		$V_{CM} = V^- + 0.1V$	●	–875	–300	0	nA
ΔI_B	Input Bias Current Shift	$V_{CM} = V^- + 0.1V$ to V^+	●		600	1750	nA
	Input Bias Current Match (Channel-to-Channel)	$V_{CM} = V^+$ (Note 5)	●	0	20	180	nA
		$V_{CM} = V^- + 0.1V$ (Note 5)	●	–180	–20	0	nA
I_{OS}	Input Offset Current	$V_{CM} = V^+$	●		15	90	nA
		$V_{CM} = V^- + 0.1V$	●		15	90	nA
ΔI_{OS}	Input Offset Current Shift	$V_{CM} = V^- + 0.1V$ to V^+	●		30	180	nA
A_{VOL}	Large-Signal Voltage Gain	$V_O = -14.5V$ to $14.5V$, $R_L = 10k$	●	900	5000		V/mV
		$V_O = -10V$ to $10V$, $R_L = 2k$	●	400	2000		V/mV
	Channel Separation	$V_O = -10V$ to $10V$, $R_L = 2k$	●	112	125		dB
CMRR	Common Mode Rejection Ratio	$V_{CM} = V^- + 0.1V$ to V^+	●	92	103		dB
	CMRR Match (Channel-to-Channel) (Note 5)	$V_{CM} = V^- + 0.1V$ to V^+	●	86	103		dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 5V$ to $\pm 15V$	●	88	103		dB
	PSRR Match (Channel-to-Channel) (Note 5)	$V_S = \pm 5V$ to $\pm 15V$	●	82	103		dB
V_{OL}	Output Voltage Swing (Low) (Note 6)	No Load	●		18	40	mV
		$I_{SINK} = 0.5mA$	●		45	90	mV
		$I_{SINK} = 10mA$	●		270	520	mV
V_{OH}	Output Voltage Swing (High) (Note 6)	No Load	●		3.5	15	mV
		$I_{SOURCE} = 0.5mA$	●		60	120	mV
		$I_{SOURCE} = 10mA$	●		480	1000	mV
I_{SC}	Short-Circuit Current		●	±12	±28		mA
I_S	Supply Current per Amplifier		●		1.9	2.8	mA
GBW	Gain-Bandwidth Product (Note 7)		●	6.1	9		MHz
SR	Slew Rate	$A_V = -1$, R_L = Open, $V_O = \pm 10V$ Measured at $V_O = \pm 5V$	●	3.4	5.3		V/µs

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the temperature range $-40°C < T_A < 85°C$. $V_S = \pm 15V$, $V_{CM} = 0V$, $V_{OUT} = 0V$, unless otherwise noted. (Note 4)

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	$V_{CM} = V^+$ $V_{CM} = V^- + 0.1V$	● ●		300 300	950 950	µV µV
V_{OS} TC	Input Offset Voltage Drift (Note 3)	 $V_{CM} = V^+$	● ●		1.0 2.0	3.5 5.0	µV/°C µV/°C
ΔV_{OS}	Input Offset Voltage Shift	$V_{CM} = V^- + 0.1V$ to V^+	●		250	850	µV
	Input Offset Voltage Match (Channel-to-Channel)	$V_{CM} = V^- + 0.1V$, V^+ (Note 5)	●		350	1800	µV
I_B	Input Bias Current	$V_{CM} = V^+$ $V_{CM} = V^- + 0.1V$	● ●	0 –1050	350 –350	1050 0	nA nA
ΔI_B	Input Bias Current Shift	$V_{CM} = V^- + 0.1V$ to V^+	●		700	2100	nA
	Input Bias Current Match (Channel-to-Channel)	$V_{CM} = V^+$ (Note 5) $V_{CM} = V^- + 0.1V$ (Note 5)	● ●	0 –200	20 –20	200 0	nA nA
I_{OS}	Input Offset Current	$V_{CM} = V^+$ $V_{CM} = V^- + 0.1V$	● ●		15 15	115 115	nA nA
ΔI_{OS}	Input Offset Current Shift	$V_{CM} = V^- + 0.1V$ to V^+	●		30	230	nA
A_{VOL}	Large-Signal Voltage Gain	$V_O = -14.5V$ to $14.5V$, $R_L = 10k$ $V_O = -10V$ to $10V$, $R_L = 2k$	● ●	800 350	5000 2000		V/mV V/mV
	Channel Separation	$V_O = -10V$ to $10V$, $R_L = 2k$	●	110	120		dB
CMRR	Common Mode Rejection Ratio	$V_{CM} = V^- + 0.1V$ to V^+	●	90	101		dB
	CMRR Match (Channel-to-Channel) (Note 5)	$V_{CM} = V^- + 0.1V$ to V^+	●	86	100		dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 5V$ to $\pm 15V$	●	88	100		dB
	PSRR Match (Channel-to-Channel) (Note 5)	$V_S = \pm 5V$ to $\pm 15V$	●	82	100		dB
V_{OL}	Output Voltage Swing (Low) (Note 6)	No Load $I_{SINK} = 0.5mA$ $I_{SINK} = 10mA$	● ● ●		25 50 275	50 100 520	mV mV mV
V_{OH}	Output Voltage Swing (High) (Note 6)	No Load $I_{SOURCE} = 0.5mA$ $I_{SOURCE} = 10mA$	● ● ●		3.5 65 500	15 120 1000	mV mV mV
I_{SC}	Short-Circuit Current		●	±10	±18		mA
I_S	Supply Current per Amplifier		●		2.0	3.0	mA
GBW	Gain-Bandwidth Product (Note 7)		●	5.8	8.5		MHz
SR	Slew Rate	$A_V = -1$, R_L = Open, $V_O = \pm 10V$, Measure at $V_O = \pm 5V$	●	3	4.75		V/µs

Note 1: Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

Note 2: A heat sink may be required to keep the junction temperature below the absolute maximum rating when the output is shorted indefinitely.

Note 3: This parameter is not 100% tested.

Note 4: The LT1498C/LT1499C are guaranteed to meet specified performance from 0°C to 70°C. The LT1498C/LT1499C are designed, characterized and expected to meet specified performance from –40°C to 85°C but are not tested or QA sampled at these temperatures. The LT1498I/LT1499I are guaranteed to meet specified performance from –40°C to 85°C.

Note 5: Matching parameters are the difference between amplifiers A and D and between B and C on the LT1499; between the two amplifiers on the LT1498.

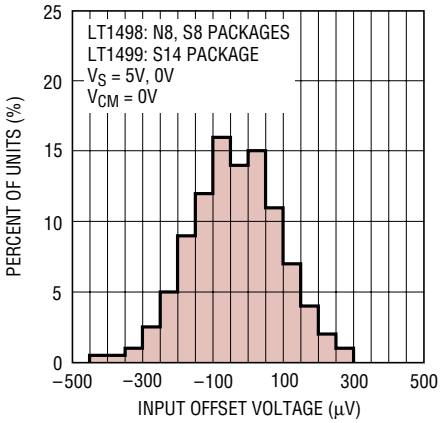
Note 6: Output voltage swings are measured between the output and power supply rails.

Note 7: $V_S = 3V$, $V_S = \pm 15V$ GBW limit guaranteed by correlation to 5V tests.

Note 8: $V_S = 3V$, $V_S = 5V$ slew rate limit guaranteed by correlation to ±15V tests.

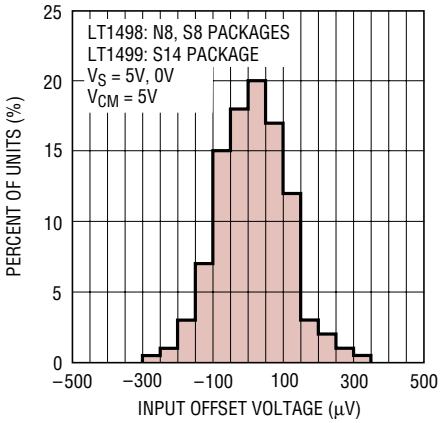
TYPICAL PERFORMANCE CHARACTERISTICS

V_{OS} Distribution, V_{CM} = 0V (PNP Stage)



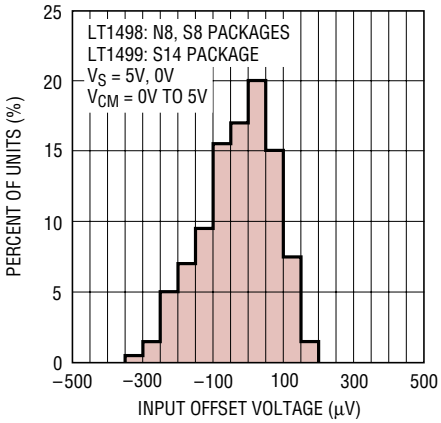
1498/99 G01

V_{OS} Distribution V_{CM} = 5V (NPN Stage)



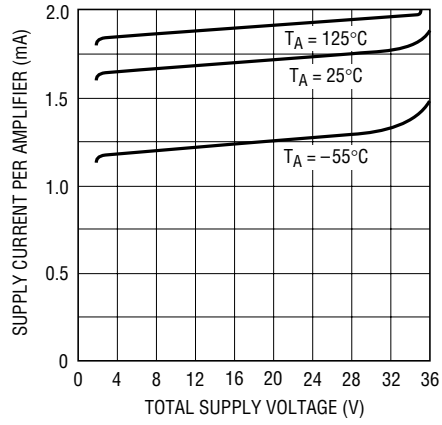
1498/99 G02

ΔV_{OS} Shift for V_{CM} = 0V to 5V



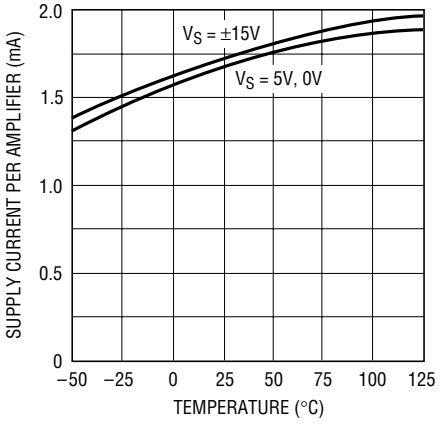
1498/99 G02

Supply Current vs Supply Voltage



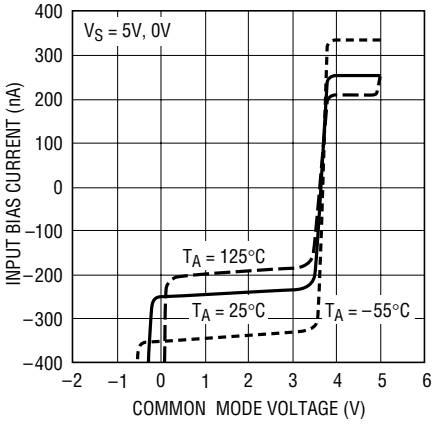
1498/99 G04

Supply Current vs Temperature



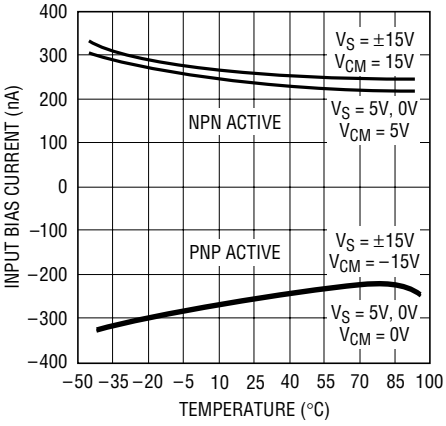
1498/99 G05

Input Bias Current vs Common Mode Voltage



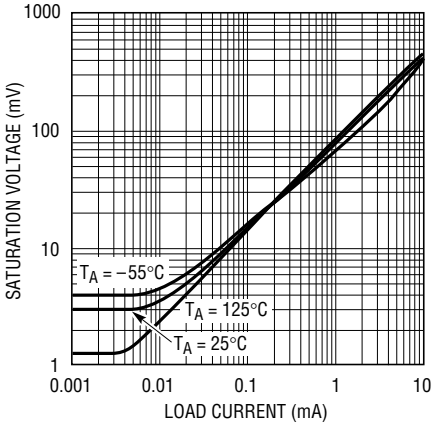
1498/99 G06

Input Bias Current vs Temperature



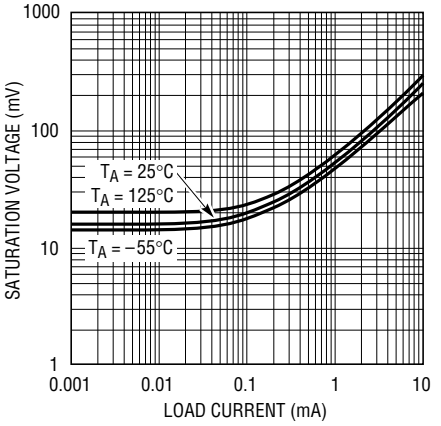
1498/99 G07

Output Saturation Voltage vs Load Current (Output High)



1498/99 G08

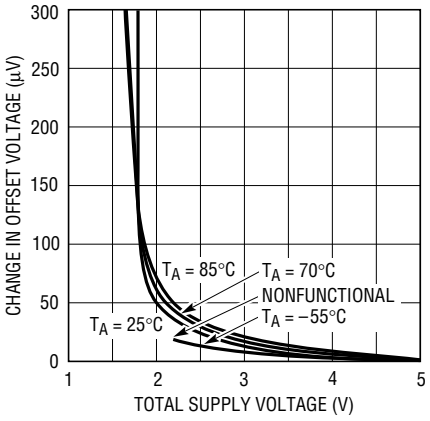
Output Saturation Voltage vs Load Current (Output Low)



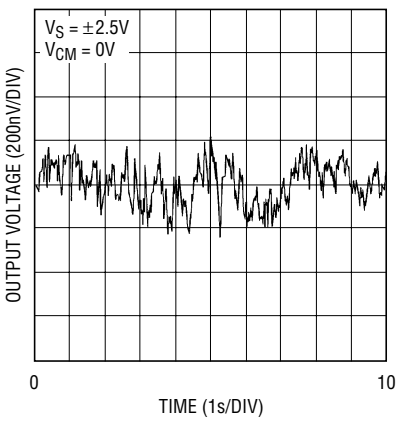
1498/99 G09

TYPICAL PERFORMANCE CHARACTERISTICS

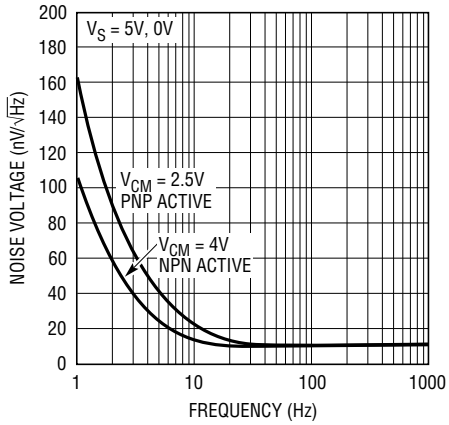
Minimum Supply Voltage



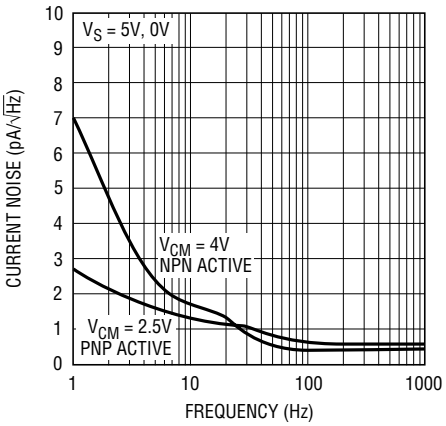
0.1Hz to 10Hz Output Voltage Noise



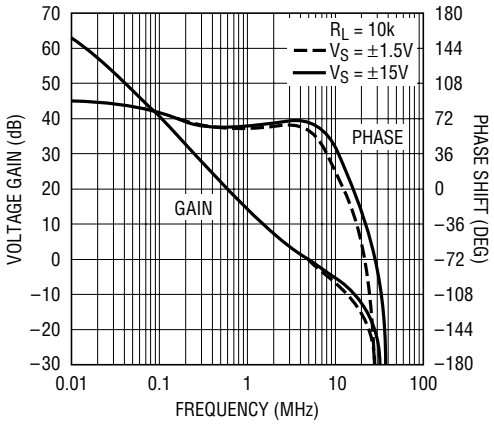
Noise Voltage Spectrum



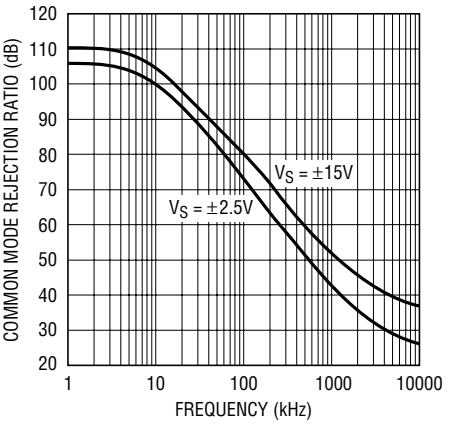
Noise Current Spectrum



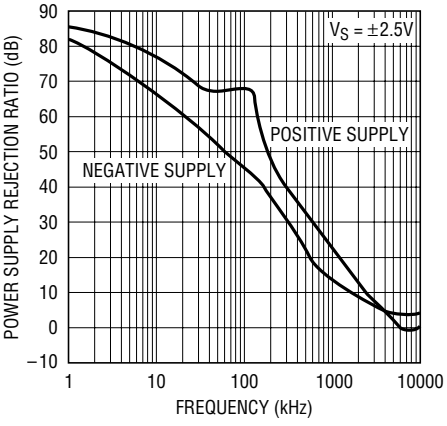
Gain and Phase vs Frequency



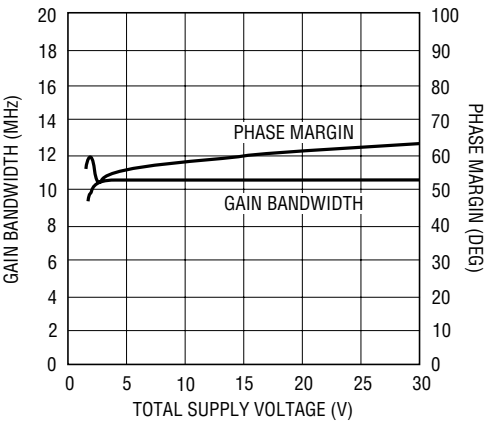
CMRR vs Frequency



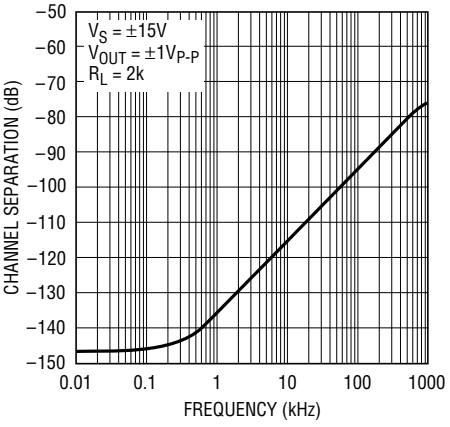
PSRR vs Frequency



Gain Bandwidth and Phase Margin vs Supply Voltage

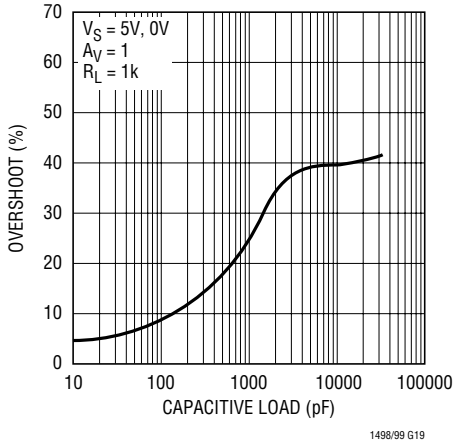


Channel Separation vs Frequency

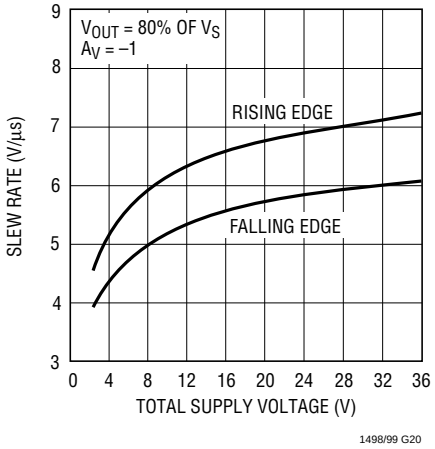


TYPICAL PERFORMANCE CHARACTERISTICS

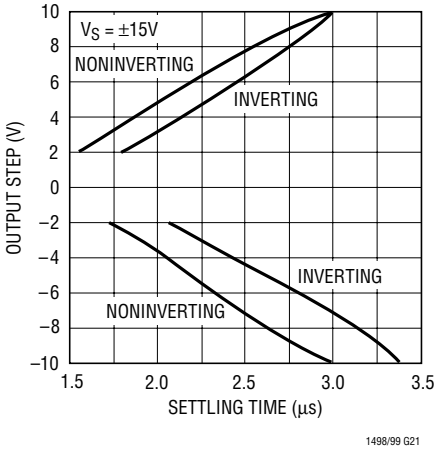
Capacitive Load Handling



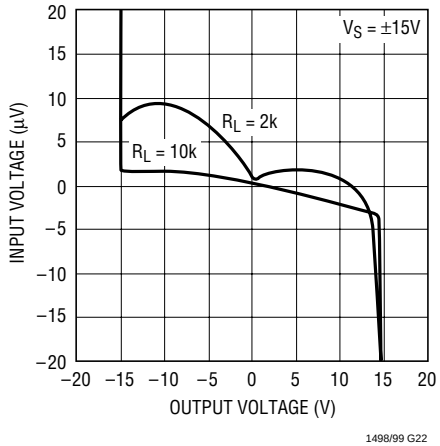
Slew Rate vs Supply Voltage



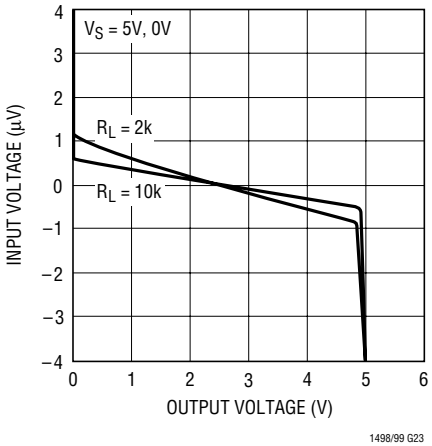
Output Step vs Settling Time to 0.01%



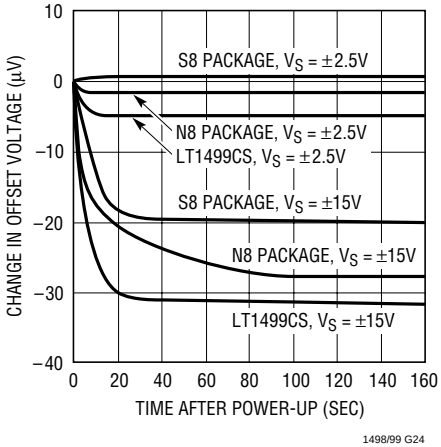
Open-Loop Gain



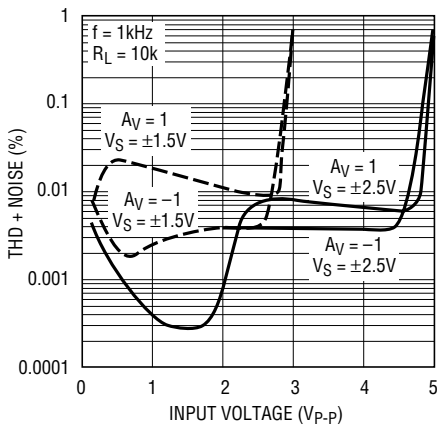
Open-Loop Gain



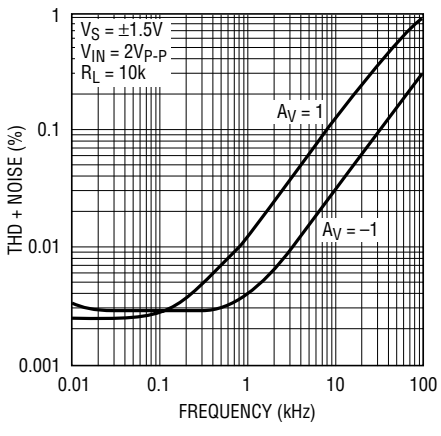
Warm-Up Drift vs Time



Total Harmonic Distortion + Noise vs Peak-to-Peak Voltage

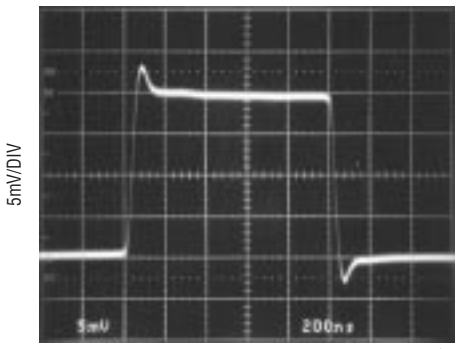


Total Harmonic Distortion + Noise vs Frequency



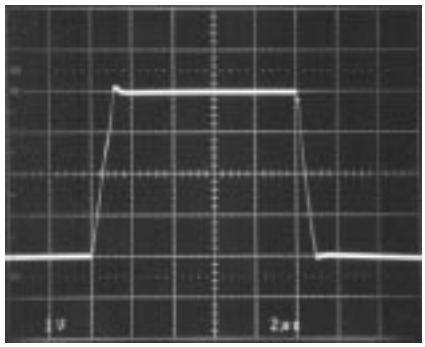
TYPICAL PERFORMANCE CHARACTERISTICS

5V Small-Signal Response



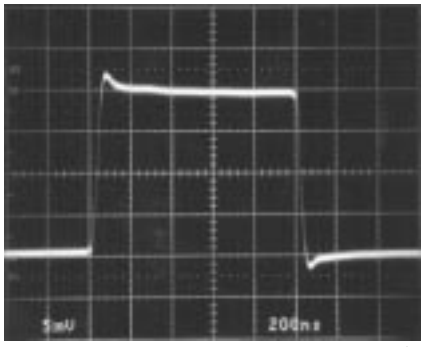
V_S = 5V 200ns/DIV
A_V = 1
V_{IN} = 20mV_{P-P} AT 50kHz
R_L = 1k

5V Large-Signal Response



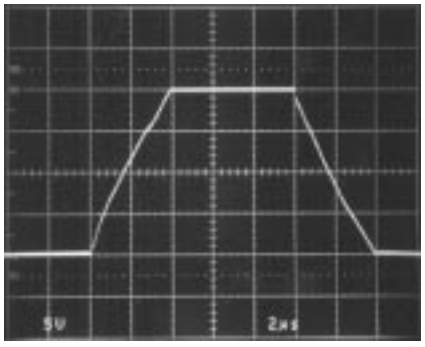
V_S = 5V 2µs/DIV
A_V = 1
V_{IN} = 4V_{P-P} AT 10kHz
R_L = 1k

±15V Small-Signal Response



V_S = ±15V 200ns/DIV
A_V = 1
V_{IN} = 20mV_{P-P} AT 50kHz
R_L = 1k

±15V Large-Signal Response



V_S = ±15V 2µs/DIV
A_V = 1
V_{IN} = 20V_{P-P} AT 10kHz
R_L = 1k

APPLICATIONS INFORMATION

Rail-to-Rail Input and Output

The LT1498/LT1499 are fully functional for an input and output signal range from the negative supply to the positive supply. Figure 1 shows a simplified schematic of the amplifier. The input stage consists of two differential amplifiers, a PNP stage (Q1/Q2) and an NPN stage (Q3/Q4) which are active over different ranges of input common mode voltage. A complementary common emitter output stage (Q14/Q15) is employed allowing the output to swing from rail-to-rail. The devices are fabricated on Linear Technology's proprietary complementary bipolar process to ensure very similar DC and AC characteristics for the output devices (Q14/Q15).

The PNP differential input pair is active for input common mode voltages, V_{CM}, between the negative supply to approximately 1.3V below the positive supply. As V_{CM} moves further toward the positive supply, the transistor Q5 will steer the tail current, I_1, to the current mirror Q6/Q7 activating the NPN differential pair, and the PNP differential pair becomes inactive for the rest of the input common mode range up to the positive supply.

The output is configured with a pair of complementary common emitter stages that enables the output to swing from rail to rail. Capacitors C1 and C2 form local feedback loops that lower the output impedance at high frequencies.

APPLICATIONS INFORMATION

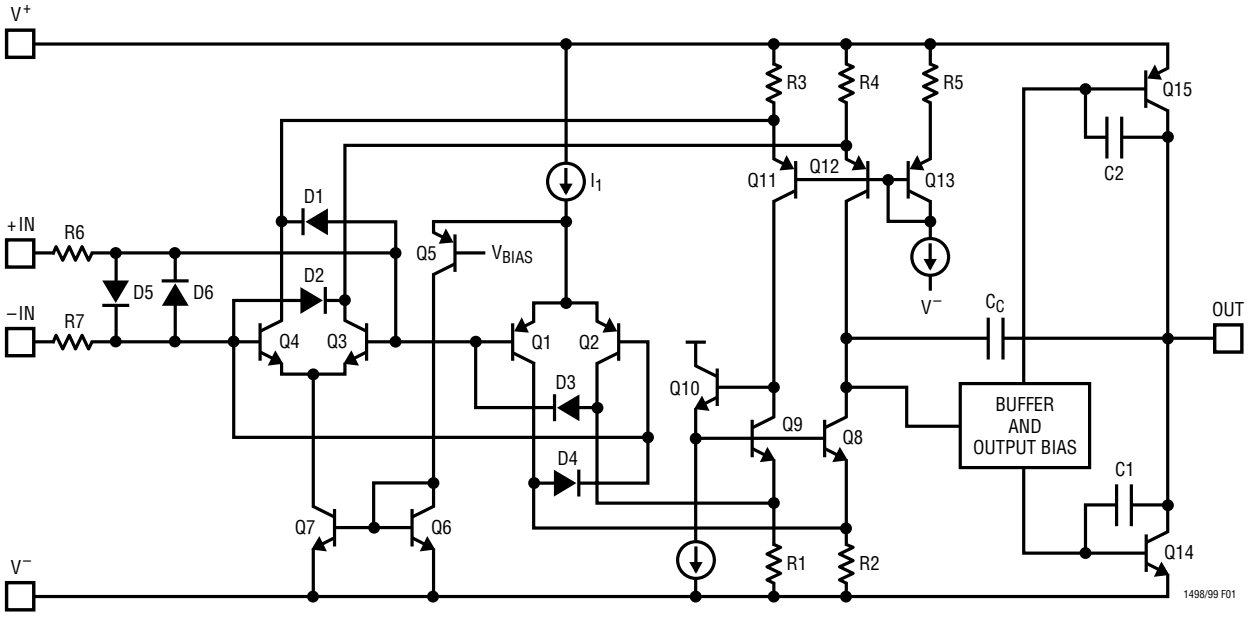


Figure 1. LT1498 Simplified Schematic Diagram

Input Offset Voltage

The offset voltage changes depending upon which input stage is active. The input offsets are random, but are trimmed to less than 475μV. To maintain the precision characteristics of the amplifier, the change of V_{OS} over the entire input common mode range (CMRR) is guaranteed to be less than 425μV on a single 5V supply.

Input Bias Current

The input bias current polarity also depends on the input common mode voltage, as described in the previous section. When the PNP differential pair is active, the input bias currents flow out of the input pins; they flow in opposite direction when the NPN input stage is active. The offset error due to input bias current can be minimized by equalizing the noninverting and inverting input source impedances. This will reduce the error since the input offset currents are much less than the input bias currents.

Overdrive Protection

To prevent the output from reversing polarity when the input voltage exceeds the power supplies, two pair of crossing diodes D1 to D4 are employed. When the input voltage exceeds either power supply by approximately 700mV, D1/D2 or D3/D4 will turn on, forcing the output to the proper polarity. For the phase reversal protection to work properly, the input current must be less than 5mA. If the amplifier is to be severely overdriven, an external resistor should be used to limit the overdrive current.

Furthermore, the LT1498/LT1499's input stages are protected by a pair of back-to-back diodes, D5/D6. When a differential voltage of more than 0.7V is applied to the inputs, these diodes will turn on, preventing the Zener breakdown of the input transistors. The current in D5/D6 should be limited to less than 10mA. Internal resistors R6 and R7 (700Ω total) limit the input current for differential input signals of 7V or less. For larger input levels, a resistor in series with either or both inputs should be used to limit the current. Worst-case differential input voltage usually occurs when the output is shorted to ground. In addition, the amplifier is protected against ESD strikes up to 3kV on all pins.

Capacitive Load

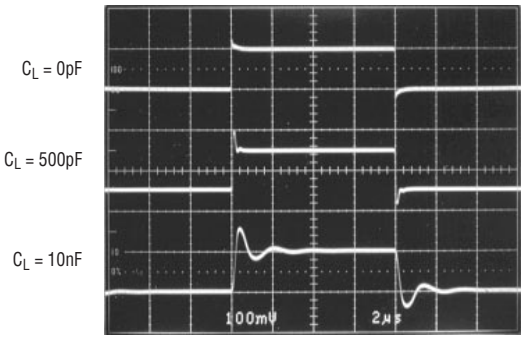
The LT1498/LT1499 are designed for ease of use. The amplifier can drive a capacitive load of more than 10nF

APPLICATIONS INFORMATION

without oscillation at unity gain. When driving a heavy capacitive load, the bandwidth is reduced to maintain stability. Figures 2a and 2b illustrate the stability of the device for small-signal and large-signal conditions with capacitive loads. Both the small-signal and large-signal transient response with a 10nF capacitive load are well behaved.

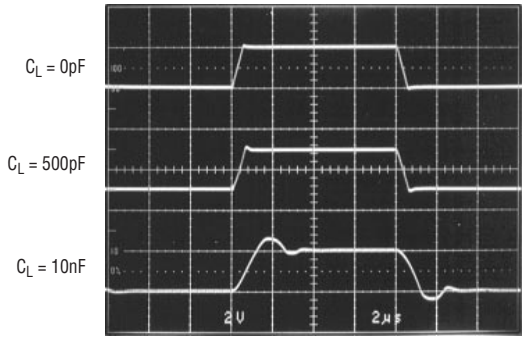
Feedback Components

To minimize the loading effect of feedback, it is possible to use the high value feedback resistors to set the gain. However, care must be taken to insure that the pole formed by the feedback resistors and the total input capacitance at the inverting input does not degrade the stability of the amplifier. For instance, the LT1498/LT1499 in a noninverting gain of 2, set with two 30k resistors, will probably oscillate with 10pF total input capacitance (5pF input capacitance + 5pF board capacitance). The amplifier has a 2.5MHz crossing frequency and a 60° phase margin at 6dB of gain. The feedback resistors and the total input capacitance create a pole at 1.06MHz that induces 67° of phase shift at 2.5MHz! The solution is simple, either lower the value of the resistors or add a feedback capacitor of 10pF of more.



V_S = 5V
A_V = 1

Figure 2a. LT1498 Small-Signal Response

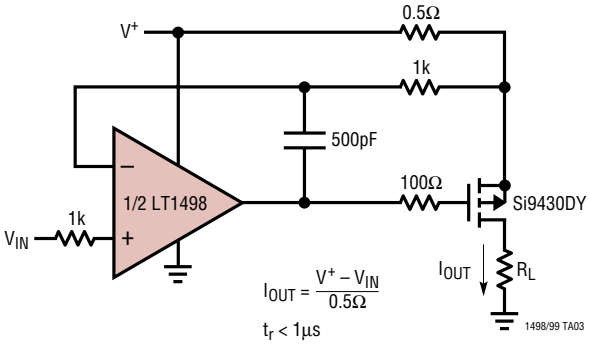


V_S = 5V
A_V = 1

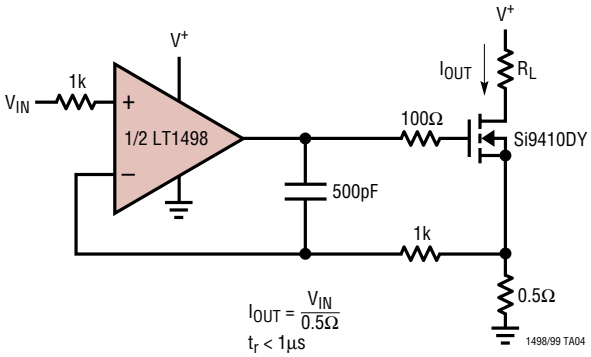
Figure 2b. LT1498 Large-Signal Response

TYPICAL APPLICATIONS

1A Voltage Controlled Current Source

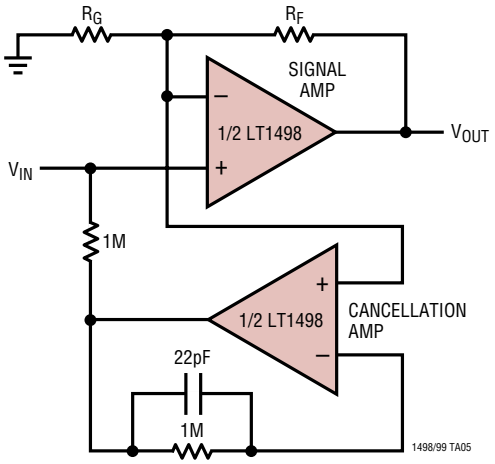


1A Voltage Controlled Current Sink



TYPICAL APPLICATIONS

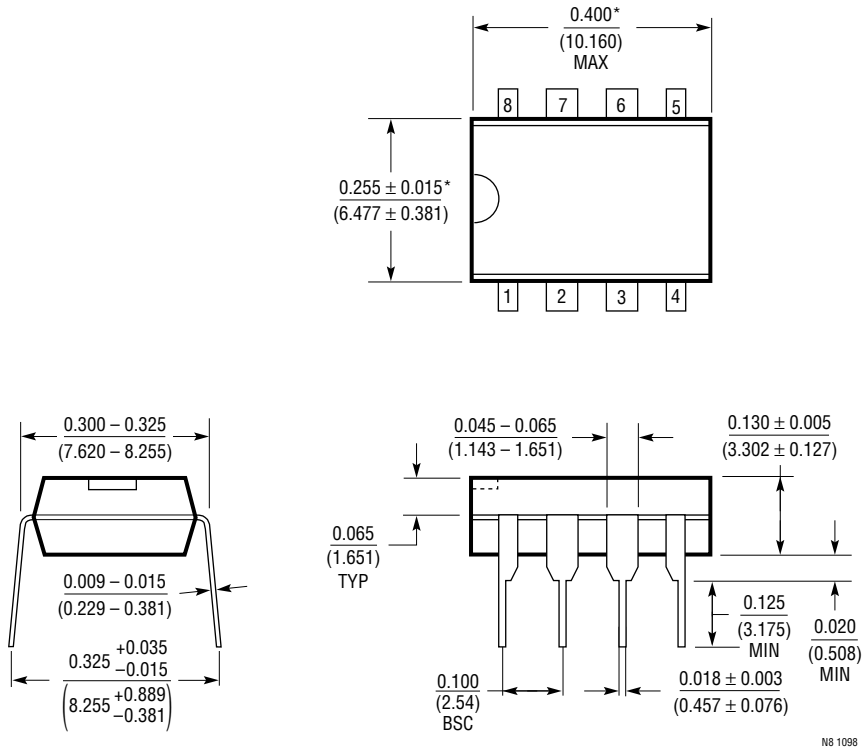
Input Bias Current Cancellation



INPUT BIAS CURRENT LESS THAN 50nA
FOR 500mV ≤ V_{IN} ≤ (V^+ – 500mV)

PACKAGE DESCRIPTION

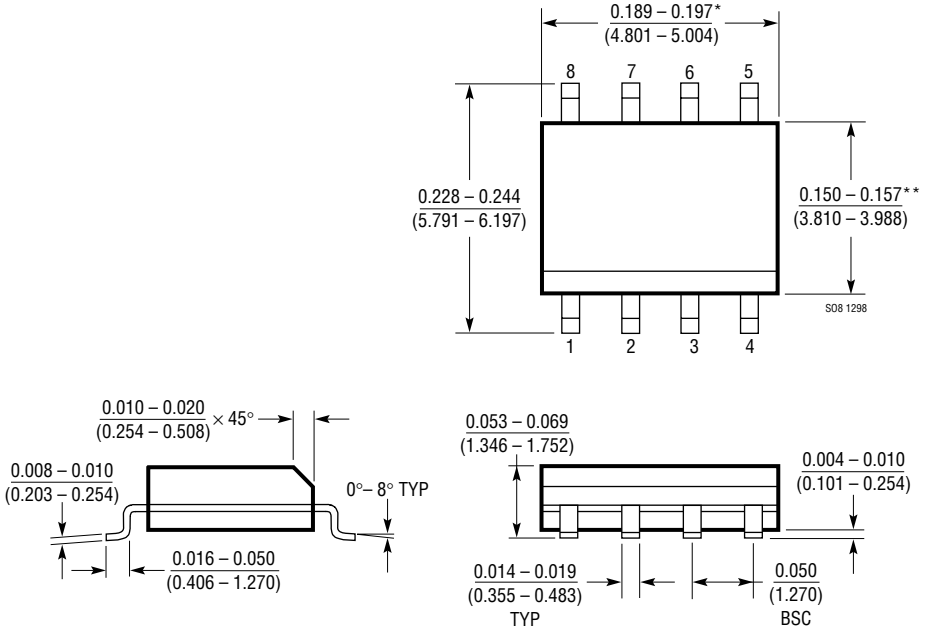
N8 Package
8-Lead PDIP (Narrow .300 Inch)
(Reference LTC DWG # 05-08-1510)



*THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS.
MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.010 INCH (0.254mm)

PACKAGE DESCRIPTION

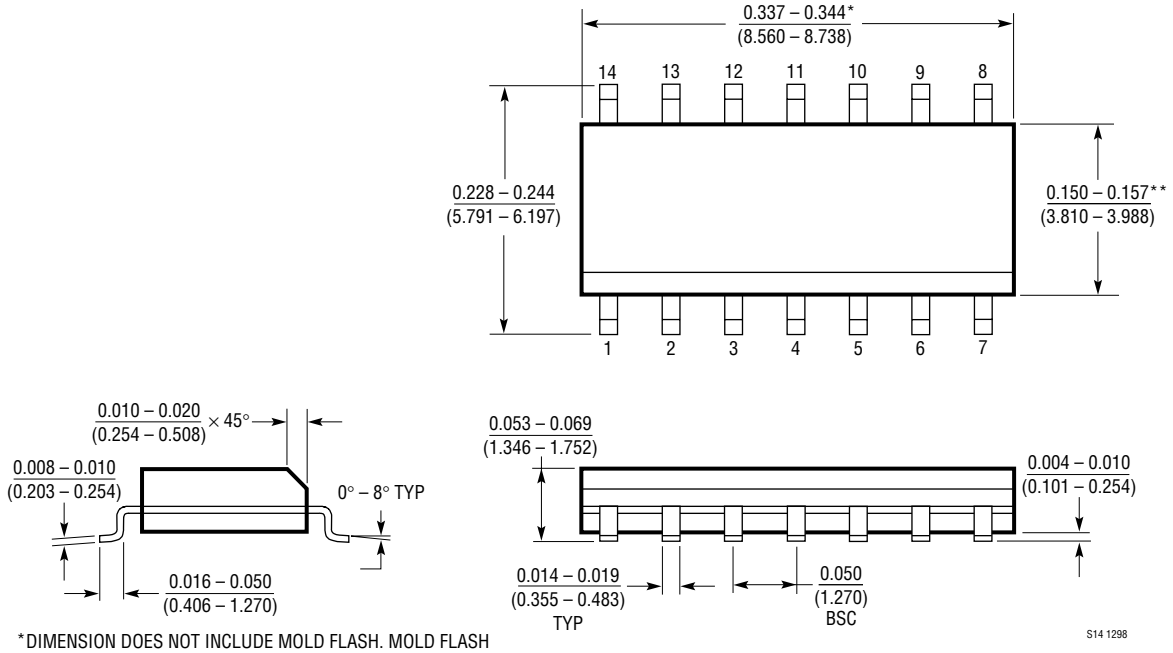
S8 Package
8-Lead Plastic Small Outline (Narrow .150 Inch)
(Reference LTC DWG # 05-08-1610)



*DIMENSION DOES NOT INCLUDE MOLD FLASH. MOLD FLASH SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE

**DIMENSION DOES NOT INCLUDE INTERLEAD FLASH. INTERLEAD FLASH SHALL NOT EXCEED 0.010" (0.254mm) PER SIDE

S Package
14-Lead Plastic Small Outline (Narrow .150 Inch)
(Reference LTC DWG # 05-08-1610)



*DIMENSION DOES NOT INCLUDE MOLD FLASH. MOLD FLASH SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE

**DIMENSION DOES NOT INCLUDE INTERLEAD FLASH. INTERLEAD FLASH SHALL NOT EXCEED 0.010" (0.254mm) PER SIDE

TYPICAL APPLICATION

Bidirectional Current Sensor

A bidirectional current sensor for battery-powered systems is shown in Figure 3. Two outputs are provided: one proportional to charge current, the other proportional to discharge current. The circuit takes advantage of the LT1498's rail-to-rail input range and its output phase reversal protection. During the charge cycle, the op amp A1 forces a voltage equal to $(I_L)(R_{SENSE})$ across R_A. This voltage is then amplified at the Charge Out by the ratio of R_B over R_A. In this mode, the output of A2 remains high, keeping Q2 off and the Discharge Out low, even though the (+) input of A2 exceeds the positive power supply. During the discharge cycle, A2 and Q2 are active and the operation is similar to the charge cycle.

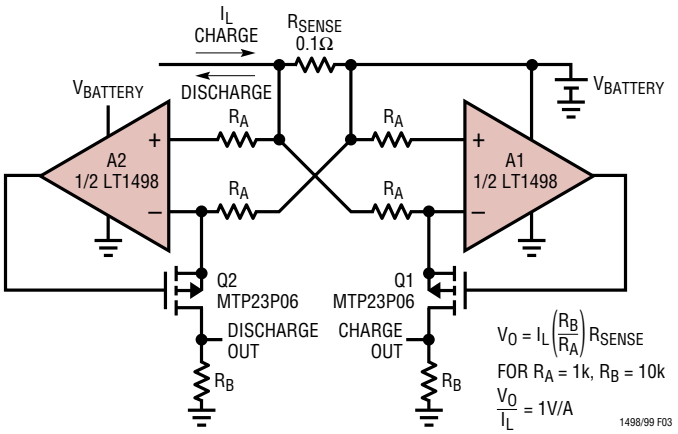


Figure 3. Bidirectional Current Sensor

RELATED PARTS

PART NUMBER	DESCRIPTON	COMMENTS
LTC®1152	Rail-to-Rail Input and Output, Zero-Drift Op Amp	High DC Accuracy, 10µV $V_{OS(MAX)}$, 100nV/°C Drift, 1MHz GBW, 1V/µs Slew Rate, Max Supply Current 2.2mA
LT1211/LT1212	Dual/Quad 14MHz, 7V/µs, Single Supply Precision Op Amps	Input Common Mode Includes Ground, 275µV $V_{OS(MAX)}$, 6µV/°C Max Drift, Max Supply Current 1.8mA per Op Amp
LT1213/LT1214	Dual/Quad 28MHz, 12V/µs, Single Supply Precision Op Amps	Input Common Mode Includes Ground, 275µV $V_{OS(MAX)}$, 6µV/°C Max Drift, Max Supply Current 3.5mA per Op Amp
LT1215/LT1216	Dual/Quad 23MHz, 50V/µs, Single Supply Precision Op Amps	Input Common Mode Includes Ground, 450µV $V_{OS(MAX)}$, Max Supply Current 6.6mA per Op Amp
LT1366/LT1367	Dual/Quad Precision, Rail-to-Rail Input and Output Op Amps	475µV $V_{OS(MAX)}$, 400kHz GBW, 0.13V/µs Slew Rate, Max Supply Current 520µA per Op Amp
LT1490/LT1491	Dual/Quad Micropower, Rail-to-Rail Input and Output Op Amps	Max Supply Current 50µA per Op Amp, 200kHz GBW, 0.07V/µs Slew Rate, Operates with Inputs 44V Above V^- Independent of V^+

Linear Technology Corporation
1630 McCarthy Blvd., Milpitas, CA 95035-7417• (408) 432-1900
FAX: (408) 434-0507 • TELEX: 499-3977 • www.linear.com
LT/TP 1101 REV A 1.5K • PRINTED IN USA